

#### DIRECT QUADRATURE MODULATOR

#### Typical Applications

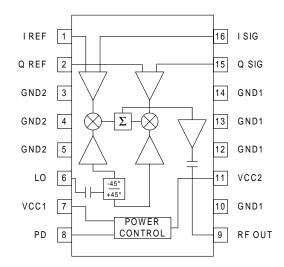
- Dual-Band CDMA Base Stations
- TDMA/TDMA-EDGE Base Stations
- GSM-EDGE/EGSM Base Stations
- W-CDMA Base Stations
- WLAN and WLL Systems
- TETRA Systems

#### **Product Description**

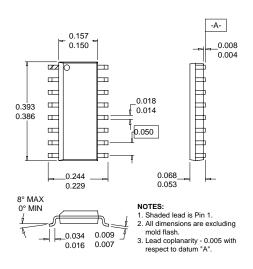
The RF2480 is a monolithic integrated quadrature modulator IC capable of universal direct modulation for high-frequency AM, PM, or compound carriers. This low-cost IC features excellent linearity, noise floor, and over-temperature carrier suppression performance. The device implements differential amplifiers for the modulation inputs, 90° carrier phase shift network, carrier limiting amplifiers, two matched double-balanced mixers, summing amplifier, and an output RF amplifier which will drive  $50\Omega$  from 800MHz to 2500MHz. Component matching is used to obtain excellent amplitude balance and phase accuracy.

Optimum Technology Matching® Applied

☐ Si BJT ☐ GaAs MESFET☐ Si Bi-CMOS☐ SiGe HBT☐ Si CMOS☐



Functional Block Diagram



Package Style: SOIC-16

#### **Features**

- Typical Carrier Suppression>35dBc over temperature with highly linear operation
- Single 5V Power Supply
- Integrated RF quadrature network
- Digitally controlled Power Down mode
- 800MHz to 2500MHz operation

#### Ordering Information

RF2480 Direct Quadrature Modulator
RF2480 PCBA Fully Assembled Evaluation Board

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## **RF2480**

#### **Absolute Maximum Ratings**

Parameter	Rating	Unit
Supply Voltage	-0.5 to +7.5	$V_{DC}$
Input LO and RF Levels	+10	dBm
Operating Ambient Temperature	-40 to +85	°C
Storage Temperature	-40 to +150	℃



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Doromotor	Specification		Unit	Condition		
Parameter	Min.	Тур.	Max.	Unit	Condition	
Carrier Input					T=25°C, V <sub>CC</sub> =5V	
Frequency Range	800		2500	MHz		
Power Level	-6		+6	dBm		
Input VSWR		4.5:1			At 900MHz unmatched	
		2:1			At 1800MHz unmatched	
		2:1			At 2500MHz unmatched	
Modulation Input						
Frequency Range	DC		250	MHz		
Reference Voltage (V <sub>REF</sub> )		3.0		V		
Maximum Modulation (I&Q)			V <sub>REF</sub> ±1.0	V		
Gain Asymmetry		0.2		dB		
Quadrature Phase Error		3		0		
Input Resistance		30		$k\Omega$		
Input Bias Current			40	μΑ		
RF Output (~800MHz)					LO=800MHz, -5dBm; SSB	
Maximum Output Power	-3	0	+2	dBm	TETRA I&Q Amplitude=2V <sub>PP</sub>	
					Over operating temperature.	
High-Linearity Output Power	-6	-5		dBm	TETRA I&Q Amplitude=1.1V <sub>PP</sub> with an	
					ACPR of -47dBc. Over operating tempera-	
					ture.	
Adjacent Channel	-47	-52		dBc	TETRA modulation applied with	
Power Rejection					P <sub>OUT</sub> =-5dBm. Over operating temperature.	
Output P1dB	+2	+3		dBm	Over operating temperature.	
IM3 Suppression	-39	-40		dBc	2kHz offset (9kHz, 11kHz) at -6dBm/tone.	
	40	50		ID.	Over operating temperature.	
IM5 Suppression	-49	-59		dBc	2kHz offset (9kHz, 11kHz) at -6dBm/tone. Over operating temperature.	
IM7 Suppression	-49	-71		dBc	2kHz offset (9kHz, 11kHz) at -6dBm/tone.	
INT Suppression	-43	-71		abc	Over operating temperature.	
Carrier Suppression	-25	-30		dBc	Unadjusted performance.	
Sideband Suppression	-25	-30		dBc	Unadjusted performance.	
Broadband Noise Floor		-150	-145	dBm/Hz	26MHz offset with TETRA signal applied	
					P <sub>OUT</sub> =-5dBm.	

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Doromotor	Specification		1114	O a malitis m		
Parameter	Min.	Тур.	Max.	Unit	Condition	
RF Output (~900MHz)					LO=880MHz, -5dBm; SSB	
Maximum Output Power	0	+4		dBm	I&Q Amplitude=2V <sub>PP</sub>	
High-Linearity Output Power		-11		dBm	I&Q Amplitude=0.325 V <sub>PP</sub>	
Carrier Suppression	50			dB	T=25°C; P <sub>OUT</sub> =-11 dBm (meets CDMA base station requirements); optimized I,Q DC offsets	
	35			dB	Over Temperature (Temperature cycled from -40°C to +85°C after optimization at T=25°C; P <sub>OUT</sub> =-11dBm)	
Sideband Suppression	50			dB	T=25°C; P <sub>OUT</sub> =-11 dBm; optimized I,Q DC offsets	
	35			dB	Over Temperature (Temperature cycled from -40°C to +85°C after optimization at T=25°C; P <sub>OUT</sub> =-11dBm)	
Output Impedance		13-j:25		Ω		
Broadband Noise Floor		-153.0		dBm/Hz	At 20 MHz offset, V <sub>CC</sub> =5V; Tied to V <sub>REF</sub> : ISIG, QSIG, IREF, and QREF.	
RF Output (~2000 MHz)					LO=2000MHz, -5dBm; SSB	
Maximum Output Power	-7	-3		dBm	I&Q Amplitude=2V <sub>PP</sub>	
High-Linearity Output Power		-17		dBm	I&Q Amplitude=0.325 V <sub>PP</sub>	
Carrier Suppression	50			dB	T=25°C; P <sub>OUT</sub> =-17dBm; optimized I,Q DC offsets	
	35			dB	Temperature cycled from -40°C to +85°C after optimization at T=25°C; P <sub>OUT</sub> =-17dBm	
Sideband Suppression	50			dB	T=25°C; P <sub>OUT</sub> =-17dBm; optimized I,Q DC offsets	
	40			dB	Temperature cycled from -40°C to +85°C after optimization at T=25°C; P <sub>OUT</sub> =-17dBm	
Output Impedance Broadband Noise Floor		58-j11 -158.0		Ω dBm/Hz	At 20 MHz offset, $V_{CC}$ =5 V; Tied to $V_{REF}$ :	
		-136.0		UDIII/I IZ	ISIG, QSIG, IREF, and QREF.	
Power Down				]		
Turn On/Off Time			100	ns		
PD Input Resistance	50		0.0	kΩ	There had a salt a sec	
Power Control "ON" Power Control "OFF"	1.0	1.2	2.8	V	Threshold voltage	
Power Supply	1.0	1.2		V	Threshold voltage	
Voltage		5		V	Specifications	
Voltage	4.5	3	6.0	V	Operating Limits	
Current	7.5	50	0.0	mA	Operating	
			25	μA	Power Down	

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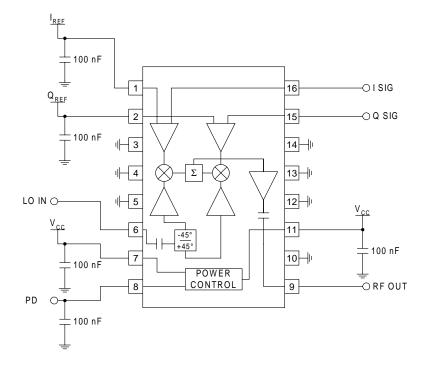
# RF2480

Pin	Function	Description	Interface Schematic
1	I REF	Reference voltage for the I mixer. This voltage should be the same as the DC voltage supplied to the I SIG pin. A voltage of 3.0 V is recommended. The SIG and REF inputs are inputs of a differential amplifier. Therefore the REF and SIG inputs are interchangeable. If swapping the I SIG and I REF pins, the Q SIG and Q REF also need to be swapped to maintain the correct phase. It is also possible to drive the SIG and REF inputs in a balanced mode. This will increase the gain.  For optimum carrier suppression, the DC voltages on I REF, Q REF, I SIG and Q SIG should be adjusted slightly to compensate for inherent undesired internal DC offsets; for optimum sideband suppression, phase and signal amplitude on IREF, Q REF, I SIG and Q SIG should be adjusted slightly to compensate for inherent undesired internal offsets. See RFMD AN0001 for more detail.	1 SIG Ο 1 REF 100 Ω 2 100 Ω 425 Ω 425 Ω
2	Q REF	Reference voltage for the Q mixer. This voltage should be the same as the DC voltage supplied to the Q SIG pin. A voltage of 3.0V is recommended. See pin 1 for more details.	Q SIG O Q REF 100 $\Omega$ 100 $\Omega$ 425 $\Omega$ 425 $\Omega$
3	GND2	Ground connection of the LO phase shift network. This pin should be connected directly to the ground plane.	
4	GND2	Same as pin 3.	
5	GND2	Same as pin 3.	
6	LO	The input of the phase shifting network. This pin has an internal DC blocking capacitor. This port is voltage driven so matching at different frequencies is not required.	ro o—vvv——  -
7	VCC1	Power supply for all circuits except the RF output stage. An external capacitor is needed if no other low frequency bypass capacitor is nearby.	
8	PD	Power Down control. When this pin is "low", all circuits are shut off. A "low" is typically 1.2 V or less at room temperature. When this pin is "high" ( $V_{CC}$ ), all circuits are operating normally. If PD is below $V_{CC}$ , output power and performance will be degraded. Operating in this region is not recommended, although it might be useful in some applications where power control is required.	PD 200 Ω
9	RF OUT	RF Output. This pin has an internal DC blocking capacitor. At some frequencies, external matching may be needed to optimize output power.	O RF OUT
10	GND3	Ground connection for the RF output stage. This pin should be connected directly to the ground plane.	
11	VCC2	Power supply for the RF output amplifier. An external capacitor is needed if no other low frequency bypass capacitor is near by.	
12	GND1	Ground connection for the LO and baseband amplifiers, and for the mixers. This pin should be connected directly to the ground plane.	
13	GND1	Same as pin 12.	
14	GND1	Same as pin 12.	

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Pin	Function	Description	Interface Schematic
15	Q SIG	Baseband input to the Q mixer. This pin is DC-coupled. Maximum output power is obtained when the input signal has a peak to peak amplitude of 2V; for highly linear operation, the input signal (and output power) must be reduced appropriately. The recommended DC level for this pin is 3.0 V. The peak minimum voltage on this pin ( $V_{REF}$ - peak modulation amplitude) should never drop below 2.0 V. The peak maximum voltage on this pin ( $V_{REF}$ + peak modulation amplitude) should never exceed 4.0 V. See pin 1 for more details.	Q SIG Ο Q REF 100 Ω \$100 Ω 425 Ω \$425 Ω
16	I SIG	Baseband input to the I mixer. This pin is DC-coupled. Maximum output power is obtained when the input signal has a peak to peak amplitude of 2V; for highly linear operation, the input signal (and output power) must be reduced appropriately. The recommended DC level for this pin is 3.0V. The peak minimum voltage on this pin (V <sub>REF</sub> - peak modulation amplitude) should never drop below 2.0V. The peak maximum voltage on this pin (V <sub>REF</sub> + peak modulation amplitude) should never exceed 4.0V. See pin 1 for more details.	1 SIG Ο 1 REF 100 Ω 100 Ω 425 Ω 425 Ω

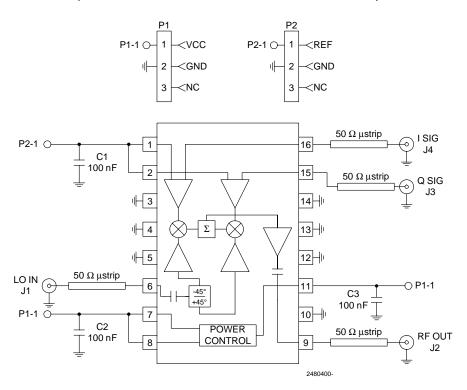
# Application Schematic DC-Coupled



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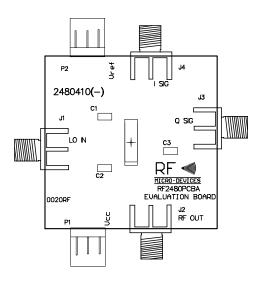
## **Evaluation Board Schematic**

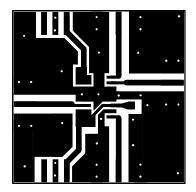
(Download Bill of Materials from www.rfmd.com.)

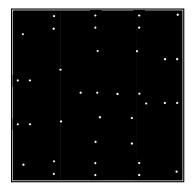


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### Evaluation Board Layout Board Size 1.510" x 1.510" Board Thickness 0.031", FR-4







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